Electron transport and heating in semiconductor devices and circuits

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- Introduction
- Device modeling
- Circuit modeling
- Numerical examples



Compact modeling approach:

- Describe devices by substituting circuits (compact models)
- Solve compact model using fitting parameters



Multiphysics modeling approach:

- Combine semiconductor and circuit models (Tischendorf 2003, Bartel/Günther 2003)
- Here: include also lattice temperature effects

Coupled device-circuit model:



- ① Electron transport: energy-transport model
- 2 Device temperature: heat equation
- ③ Circuit elements: circuit model
- ④ Heating in circuit elements: thermal network model

Energy-transport equations

Derivation from Boltzmann eq. with dominant elastic scattering (Stratton 1962, Ben Abdallah/Degond 1996)

Balance equations:

$$\partial_t n - \operatorname{div} J_n = 0$$

$$\frac{3}{2}\partial_t(nT) - \operatorname{div} S_n = -J_n \cdot \nabla V + \frac{3}{2} \frac{n(T-T_0)}{\tau}$$

Current equations:

 $J_n = \nabla(\mu_n(T_L)T_Ln) - \mu_n(T_L)T_L\frac{n}{T}\nabla V)$ $S_n = \frac{3}{2}(\nabla(\mu_n(T_L)T_LnT) - \mu_n(T_L)T_Ln\nabla V)$

- T electron temperature, T_L lattice temperature
- Mobility $\mu_n(T_L) = \mu_0(T_0/T_L)^{\alpha}$: α depends on elastic scattering model (Degond/A.J./Pietra 2000)

Energy-transport equations: electrons and holes $\partial_t n - \operatorname{div} J_n = R(n, p), \quad J_n = \nabla n - \frac{n}{T} \nabla V$ $\frac{3}{2} \partial_t (nT) - \operatorname{div} S = -J_n \cdot \nabla V + \frac{3}{2} \frac{n(T-T_0)}{\tau}$ $S = \frac{3}{2} (\nabla (nT) - n \nabla V)$ $\lambda^2 \Delta V = n - p - C(x)$ $\partial_t p - \operatorname{div} J_p = R(n, p), \quad J_p = \nabla p + p \nabla V$

Numerical difficulties:

- Convection dominance due to high electric fields
- \bullet Positivity preservation of $n,\ p$ and $T,\ current\ conservation$
- Iterative scheme for nonlinear discrete system

Discretization of stationary equations

$$-\operatorname{div} J + cn = f, \quad J = \nabla n - \frac{n}{T} \nabla V \quad \text{in } \Omega = \bigcup_{i} K_{i}$$

Formulation in extensive variables n, $\frac{3}{2}nT$:

- Slotboom var. $z = e^{-V/T}n$, $T|_{K_i} = const. \Rightarrow J = e^{V/T}\nabla z$
- Mixed finite elements: $z_h \in L^2$ piecewise constant, $J_h \in H_{loc}(div)$ Marini-Pietra (P_2 with 3 DOF) \rightarrow M-matrix
- Decoupling Gummel method (Newton in Poisson equation)

Formulation in dual entropy variables $u_1 = \frac{\mu - V}{T}$, $u_2 = -\frac{1}{T}$:

- Chem. potential μ : $n = T^{3/2}e^{\mu/T} \Rightarrow$ eliminates convection
- Mixed finite elements: $z_h \in L^2$ piecewise constant, $J_h \in H_{\text{loc}}(\text{div})$ Raviart-Thomas \rightarrow M-matrix
- Full Newton or Gummel with vector extrapolation

$\textcircled{1} \quad \textbf{Device modeling}$

3D Single-Gate MESFET (Gadau/A.J. 2008)



1 Device modeling



(2) Lattice heat modeling

- Derivation from first thermodynamic principles (Wachutka 1990, Albinus et al. 2002, Brunk/A.J. 2008)
- Define internal energy u and energy flux J_u :

 $u = ext{electric} + ext{lattice} + ext{conduction-band} + ext{thermal energies}$

 $J_u = \text{displacement} + \text{Fourier} + \text{particle} + \text{dissipated-power} + \text{band-energy fluxes}$

• Conservation equation $\partial_t u - \operatorname{div} J_u = 0 \Rightarrow$ heat equation $\rho_L c_L \partial_t T_L - \operatorname{div} (\kappa_L \nabla T_L) = H$

 $H = -\frac{3}{2} \frac{n(T - T_L)}{\tau} + R(n, p)(E_c + \frac{3}{2}T) - S_L(T_L - T_0)$

material density ρ_L , lattice heat capacity c_L , radiation $S_L(T_L - T_0)$, conduction-band energy E_c

• Coupling to energy-transport through energy relaxation

(2) Lattice heat modeling

1D bipolar diode: forward bias 1.5 V (Brunk/A.J. 2008) Electron temperature Lattice temperature



- Temperature minimum around junction
- Almost constant lattice temperature due to high heat conductivity
- Lattice heating decreases current density

- Kirchhoff's current law: Ai = 0A: matrix of node-to-branch relations, i: branch currents
- Kirchhoff's voltage law: $v = A^{\top} e$ v: branch voltages, e: node potentials



• Current-voltage characteristics:

$$v_R = Ri_R, \quad i_C = C \frac{dv_C}{dt}, \quad v_L = L \frac{di_L}{dt}$$

Modified nodal analysis: source current i_s , voltage v_s

$$egin{aligned} &A_C C A_C^ op rac{de}{dt} + A_R R^{-1} A_R^ op e + A_L i_L + A_V i_V + A_I i_s = 0 \ &L rac{di_L}{dt} - A_L^ op e = 0, \quad A_V^ op e = v_s \end{aligned}$$

- \bullet Differential-algebraic equations of index 1 or 2
- Numerical methods: Rosenbrock-Wanner (Günther), BDF

Heating of thermal nodes:

- Thermal nodes = resistors, contact nodes...
- ODE for node temperature \widehat{T} :

 $\widehat{M}\frac{d}{dt}\widehat{T} =$ source terms + radiation + dissipation power

• Dissipation power = current \times electric potential

Heating of distributed elements:

- Distributed elements = electric lines, 1D devices...
- PDE for distributed temperature T^d :

 $M\partial_t T^d = \partial_x(\kappa_L \partial_x T^d) + radiation + dissipation power$

Coupling to circuit and device models:

- Electro-thermal coupling: dissipation powers
- Thermal-device coupling: through boundary conditions

Model overview

Complete electro-thermal circuit-device model



- → Nonlinear partial differential-algebraic equations (PDAE)
- Discretization: BDF-2 and Marini-Pietra finite elements
- Iteration: Gummel-type inner loop for nonisothermal semiconductor model, fixed-point-type outer loop



Lattice temperature

- Used for entrance protection
- Output voltage should be less than 4 V
- \bullet Input frequency of 10 GHz



Summary

Novelties:

- Modeling of nonisothermal semiconductor-circuit system
- Numerical solution of complete nonlinear system

Final result:

Lattice heating may lead to undesired large output signal in clipper circuit